

# DESIGN AND IMPLEMENTATION OF AN OBSERVER-BASED SOFT SENSOR FOR A HEAT EXCHANGER

## DISEÑO E IMPLEMENTACIÓN DE UN SENSOR VIRTUAL BASADO EN OBSERVADOR PARA UN INTERCAMBIADOR DE CALOR

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**ABSTRACT:** The objective of this work is to describe step-by-step how to implement an observer-based soft sensor in order to estimate process variables for which a hardware sensor is not available. The design and implementation procedure is illustrated by applying it to a counter-flow double-pipe heat exchanger. The approach used to design the nonlinear observer is based on a simplified mathematical model of the process. Numerical simulations and experiments were performed in a bench-scale pilot plant in order to validate the proposed scheme.

**KEYWORDS:** Soft sensor, nonlinear observer, heat exchanger.

**RESUMEN:** El objetivo de este trabajo es describir paso a paso la implementación de un sensor basado en observador para estimar las variables de un proceso para el cual no existe la disponibilidad de un sensor físico. El procedimiento de diseño e implementación se ilustra mediante su aplicación a un intercambiador de calor de doble tubo con flujos a contracorriente. El enfoque empleado para diseñar el observador no lineal se basa en un modelo matemático simplificado del proceso. Se desarrollaron simulaciones numéricas y posteriormente pruebas experimentales en una planta piloto (intercambiador de calor) con el fin de validar el esquema propuesto.

**PALABRAS CLAVE:** Sensor virtual, observador no lineal, intercambiador de calor.

### 1. INTRODUCTION

SOFT sensors are an alternative to estimate process variables [1-2]: they can be used to replace costly sensors. A soft sensor can be defined as the association between a hardware sensor and an estimation algorithm as shown in Fig. 1 and studied by Farza *et. al.* in [3].

The estimation algorithm (also called state observer in this context) is the software part which performs the on-line estimation of the process variables using the available measurements. Therefore, a state

observer is the main mathematical tool which makes the conception of a soft sensor possible.

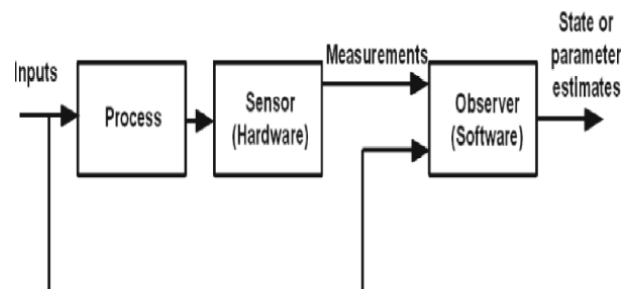


Figure 1. Principle of an observer-based soft sensor

State observers have been studied since 1960. The works developed by Kalman in [4] and by Luenberger in [5] solve the problem of state estimation for linear systems [26]. The property of observability, characterized by the rank condition guarantees the possibility of indeed designing an observer. In the nonlinear case, the observability depends on the input of the system. In this case, the observability of a nonlinear system does not exclude the existence of singular inputs (inputs for which two distinct initial states cannot be distinguished by using the known measured output).

There are several approaches to designing nonlinear observers; these depend on the specific structure of the system to be observed [6-9, 27]. For instance, in Gauthier *et. al.* [6] high-gain observers have been used to estimate the states of control-affine nonlinear systems; Targui *et. al.* [7] proposed a nonlinear observer for a particular class of nonlinear systems which have a triangular structure; Besançon *et. al.* [8] present in their work an adaptive observer for state-affine nonlinear systems; while Gauthier and Kupka [9] proposed an observer for systems having the nonlinear general form  $dx/dt = f(x)$ .

The main purpose of this paper is to describe step-by-step how to implement a nonlinear observer in order to develop a soft sensor for a heat exchanger. This kind of real-time control laboratory allows both, process engineers and students of control systems to visualize the fact that a soft sensor represents a viable alternative to avoid unnecessary, expensive instrumentation in a process.

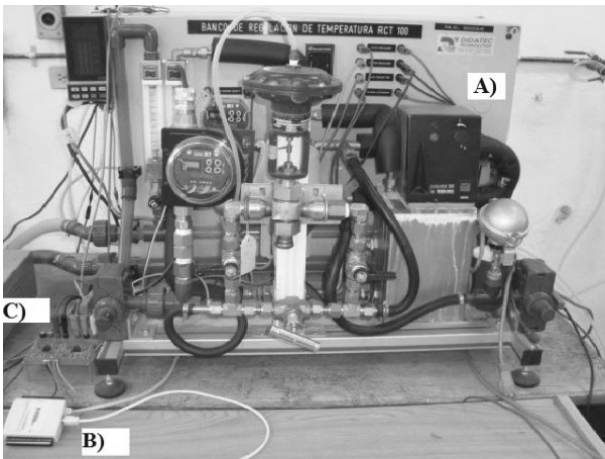


Figure 2. Counter-flow double pipe heat exchanger

## 2. PROBLEM STATEMENT

A double-pipe heat exchanger, Fig.2 (formed by two concentric tubes) can be operated in parallel (the fluids flow in the same direction through the inner and the outer tubes) or in counter-flow mode (the fluids flow in opposite directions). In this study, the second case is considered.

The main process variables of this kind of processes are: the inlet temperature of the fluid in the hot side  $T_{hi}$ , the inlet temperature of the fluid in the cold side  $T_{ci}$ , the outlet temperature of the fluid in the hot side  $T_{ho}$ , the outlet temperature of the fluid in the cold side  $T_{co}$ , the flow rate in the hot side  $v_h$  and the flow rate in the cold side  $v_c$ .

The instrumentation diagram of a double-pipe heat exchanger pilot-plant developed by Didatec Technologies is shown in Fig. 3. The plant operates as a water-cooling process: the hot water flows through the inner tube and the cooling water flows in the shell (the external tube). The pilot-plant is equipped with the following instruments:  $T_{ci}$  and  $T_{co}$  are measured via two SIKA® glass thermometers (TI1 and TI2 respectively);  $T_{co}$  and  $T_{hi}$  are measured via two Engelhard Pyro-Controle Pt-100 temperature transmitters (TT1 and TT2 respectively);  $v_c$  and  $v_h$  are measured by means of two Platon variable section flowmeters (FI1 and FI2 respectively).

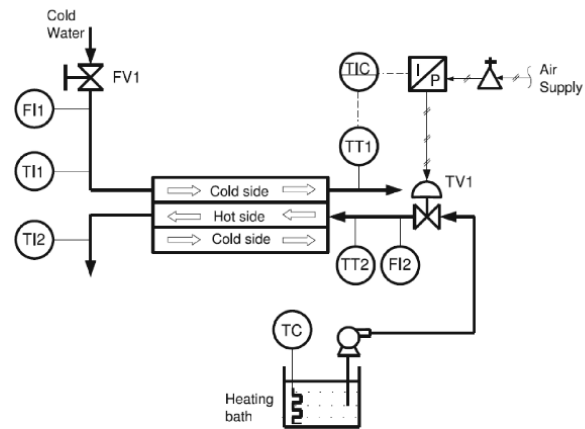


Figure 3. Instrumentation diagram of the counter-flow double-pipe heat exchanger

Let's suppose that a state-feedback control law is needed to control the outlet temperatures [10]. It is well known that this kind of control requires

the on-line measurement of the states of the whole process. In the Didatec Technologies pilot-plant this is not possible because the state variable  $T_{ho}$  is not measured electronically; so, it cannot be used for feedback control purposes. Another useful application is a remote monitoring system for whole process states. In the current configuration, the operator of the plant would have to move to the place where the thermometer is installed to be able to take the corresponding measurement.

These two problems (control and monitoring) can be solved by designing a soft sensor. The approach used to design the soft sensor is described as follows:

S1: Select or develop an adequate mathematical model of the process.

S2: Select an adequate observer depending on the model structure.

S3: Design and program the observer in order to make numerical simulations tests.

S4: Implement the observer in a computer-based system following the scheme shown in Fig.1.

The above steps of the approach are developed in the following sections.

### 3. STEP 1: HEAT EXCHANGER MATHEMATICAL MODEL

Distributed parameter models are those that best fit to the nature of heat exchangers [11], however this kind of model is often difficult to analyze or to use with feedback [10]. One of the models which can be used for control purposes is presented by Fazlur and Devanathan [12].

With this approach, the heat exchanger model can be seen as a gray box in which some lumped parameters are considered in order to simplify the whole model. Zavala-Río and Santiesteban-Cos [13] demonstrated the qualitative equivalence between the distributed-parameter models and the lumped models. Specifically, three aspects were proven to be shared by both models: existence and uniqueness of solutions, equilibrium states, and stability properties.

The authors in [14], [15] used lumped-parameter models in order to conceive a fault diagnosis system for heat exchangers. In both cases the results obtained using models considering a spatial discretization of the process were acceptable.

The heat exchanger can be divided into small elements called cells, every cell consisting of two stirred tanks as shown in Fig. 4. The model equations for a single cell are deduced from energy and mass balances for the cold and hot side.

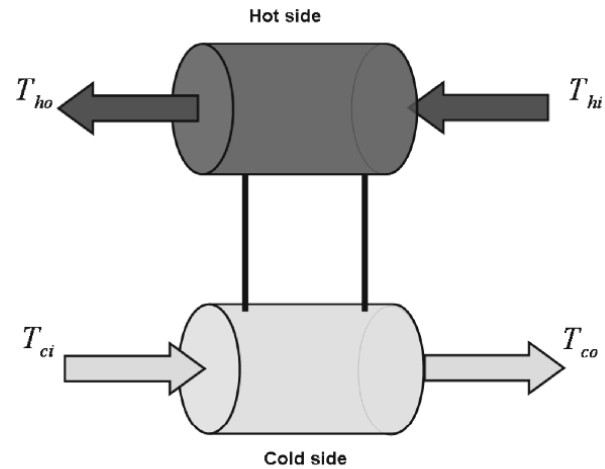


Figure 4. One-cell representation of a double-pipe heat exchanger

Before presenting the mathematical model, the following assumptions should be introduced:

A1: Adiabatic operation.

A2: The inlet temperatures  $T_{ci}$  and  $T_{hi}$  are known.

A3: The inlet temperatures  $T_{ci}$  and  $T_{hi}$  are constant.

A4: The physical and chemical properties of the fluids are constant.

A5: The global heat transfer coefficient  $U$  is constant.

Under the assumptions A1-A4, and considering an energy balance law for every cell, the single-cell heat exchanger model is given by:

$$\begin{cases} \dot{T}_{co} = \frac{2}{V_c} \left[ v_c (T_{ci} - T_{co}) + \frac{UA\Delta T}{C_{P_c} \rho_c} \right] \\ \dot{T}_{ho} = \frac{2}{V_h} \left[ v_h (T_{hi} - T_{ho}) - \frac{UA\Delta T}{C_{P_h} \rho_h} \right] \end{cases} \quad (1)$$

A complete discussion about modeling heat exchangers can be found in [12, 16, 17], where  $\Delta T$  is the (mean) temperature difference among the fluids. Basically, there are three approaches proposed in the literature for  $\Delta T$ :

i) the temperature difference  $\Delta T = T_{ho} - T_{co}$ .

ii) the arithmetic mean temperature difference (AMTD) defined as

$$\Delta T = \frac{1}{2}[(T_{ho} - T_{ci}) + (T_{hi} - T_{co})]$$

iii) the logarithmic mean temperature difference (LMTD) defined as

$$\Delta T = \frac{[(T_{ho} - T_{ci}) - (T_{hi} - T_{co})]}{\ln[(T_{ho} - T_{ci}) / (T_{hi} - T_{co})]}$$

See [18] for further information.

Zavala-Río and Santiesteban-Cos [13] proved that the use of the logarithmic mean temperature difference approach (also known as the LMTD driving-force) provides reliable dynamic representations for heat exchangers especially in cases where it is not the quantitative solutions but the qualitative behavior that is important. The LMTD driving-force is expressed in a simplified form as

$$\Delta T = \left( \frac{(\Delta T_2 - \Delta T_1)}{\ln(\Delta T_2 / \Delta T_1)} \right) \quad (2)$$

Where  $\Delta T_1 = (T_{hi} - T_{co})$  and  $\Delta T_2 = (T_{ho} - T_{ci})$

Equation (2) has a numeric indetermination when  $\Delta T_1 = \Delta T_2$  generally in the initial condition  $\Delta T_0$  for this reason a modified LMTD model is introduced to express the temperature difference  $\Delta T$ :

$$\Delta T = \begin{cases} \left( \frac{(\Delta T_2 - \Delta T_1)}{\ln(\Delta T_2 / \Delta T_1)} \right) & \Delta T_2 \neq \Delta T_1 \\ \Delta T_0 & \Delta T_2 = \Delta T_1 \end{cases} \quad (3)$$

The analytical properties of the LMTD model in (3) are reported in [13].

#### 4. STEP 2: OBSERVER SELECTION

The most-widely known observers in the literature are the extended Kalman observers (also known as extended Kalman filters) [1, 19] and the high-gain observers [6, 20]. The most common approach using the extended Kalman observer for nonlinear systems is to linearize the model. Although the extended

Kalman observers have been used successfully for several processes (see for instance [21-22]), it is well known that technically they are difficult to implement and difficult to tune, this is because a Riccati equation must be solved and a numerical instability can easily arise due to the accumulated errors [23].

High-gain observers work either for autonomous systems or for nonlinear systems that are observable for each input. One of the main features of these observers is that they are easy to implement, because the observer gain is obtained from an algebraic Lyapunov equation and is simple to compute. High-gain observers have been applied to several kinds of processes, i.e. polymerization reactors [24], distillation columns [7], and chemical reactors [25]. For these reasons (implementation and tuning facilities), in the following sections, the high-gain observer is used to design the soft sensor for the heat exchanger.

#### 4.1 The high-gain observer

The control-affine nonlinear system is given by:

$$\begin{cases} \dot{x} = f(x) + \sum_{i=1}^m u_i g_i(x) \\ y = h(x) \end{cases} \quad (4)$$

where  $x \in \mathfrak{R}$ ,  $u_i \in \mathfrak{R}$ ,  $y \in \mathfrak{R}$ ,  $f(x) \in \mathfrak{R}^n$  and  $g_i \in \mathfrak{R}^n$ . Assume that the system in (4) is observable and consider the nonlinear change of coordinates of  $z = \Phi(x)$  where  $\Phi(x) = (h(x) L_f h(x) \dots L_f^{n-1} h(x))^T$ . The notation  $L_f(\cdot)$  represents the Lie derivative of a real valued function along  $f(x)$ . By definition

$$L_f(\cdot) h(x) = \sum_{i=1}^n \frac{\partial h(x)}{\partial x_i} f_i(x) \quad (5)$$

the transformation  $\Phi(x)$  determines a system which takes the form

$$\begin{cases} \dot{z} = Az + \psi(z) + \sum_{i=1}^m u_i \phi_i(z) \\ y = Cz \end{cases} \quad (6)$$

$$\text{where } A = \begin{bmatrix} 0 & 1 & 0 & 0 \\ \vdots & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \\ 0 & \dots & 0 & 0 \end{bmatrix}, \psi(z) = \begin{bmatrix} 0 \\ \vdots \\ 0 \\ \phi_n(z) \end{bmatrix}$$

and  $C = [1 \ 0 \ \dots \ 0]$ .

The elements of  $\phi_i(z)$  are

$$\begin{aligned} \phi_1(z) &= \phi_1(z_1) \\ \phi_2(z) &= \phi_1(z_1, z_2) \\ &\vdots \\ \phi_n(z) &= \phi_1(z_1, \dots, z_{n-1}) \end{aligned}$$

For the system in the form (6), the authors in [6] have proposed an exponential observer given by

$$\begin{aligned} \dot{\hat{Z}} &= A\hat{z} + \psi(\hat{z}) + \sum_{i=1}^m u_i \phi_i \hat{z} \\ -S_\theta^{-1} C^T (C\hat{z} - y) \end{aligned} \quad (7)$$

where  $S_\theta \in \mathfrak{R}^{n \times n}$  is the symmetric positive definite matrix which is the unique solution of the Lyapunov equation

$$\theta S_\theta + A^T S_\theta + S_\theta A = C^T C \quad (8)$$

$\theta > 0$  which is the tuning parameter of the observer. The elements of  $S_\theta$  are

$$(S_\theta)_{ij} = S_{ij} / \theta^{i+j-1} \quad \text{for } 1 \leq i, j \leq n \quad (9)$$

where  $S_{ij}$  is a known combinatory coefficient. For instance, considering a second order system, the matrix  $S_\theta$  is:

$$S_\theta = \begin{bmatrix} \frac{1}{\theta} & -\frac{1}{\theta^2} \\ -\frac{1}{\theta^2} & \frac{2}{\theta^3} \end{bmatrix} \quad (10)$$

Transforming the system given in (7) back into the original coordinates, the observer gets the following form:

$$\begin{aligned} \dot{\hat{x}} &= f(\hat{x}) + \sum_{i=1}^m u_i g_i(\hat{x}) \\ -\left[ \frac{\partial \Phi(\hat{x})}{\partial(\hat{x})} \right]^{-1} S_\theta C^T [C\hat{x} - y] \\ \hat{y} &= C\hat{x} \end{aligned} \quad (11)$$

This observer is known in the literature as the *high-gain observer*. The authors in [6] have demonstrated that if the value of the tuning parameter  $\theta$  is selected high enough, then the estimation error  $\|\hat{x}(t) - x(t)\|$  converges exponentially towards zero.

### 5. STEP 3: NUMERICAL SOLUTIONS

The heat exchanger model in (1) can be written in the following matrix form:

$$\begin{bmatrix} \dot{T}_{co} \\ \dot{T}_{ho} \end{bmatrix} = \begin{bmatrix} k_c \Delta T \\ -k_h \Delta T \end{bmatrix} + \begin{bmatrix} \frac{2}{V_c} (T_{ci} - T_{co}) \\ \frac{2}{V_h} (T_{hi} - T_{ho}) \end{bmatrix} \begin{bmatrix} v_c \\ v_h \end{bmatrix}$$

$$y = T_{co} \quad (12)$$

Where  $k_c = \frac{2UA}{(C_p \rho_c V_c)}$  and  $k_h = \frac{2UA}{(C_p \rho_h V_h)}$ . The model in (12) has the same form of the system in (4). Hence, it is possible to design a high-order observer (13) of the form given in (11) as follows:

$$\begin{bmatrix} \dot{T}_{co} \\ \dot{T}_{ho} \end{bmatrix} = \begin{bmatrix} k_c \Delta \hat{T} \\ -k_h \Delta \hat{T} \end{bmatrix} + \begin{bmatrix} \frac{2}{V_c} (T_{ci} - \hat{T}_{co}) \\ \frac{2}{V_h} (T_{hi} - \hat{T}_{ho}) \end{bmatrix} \begin{bmatrix} v_c \\ v_h \end{bmatrix}$$

$$+ \left. \frac{\partial \Phi(\hat{x})^{-1}}{\partial(\hat{x})} \right|_{x=(\hat{T}_{co}, \hat{T}_{ho})^T} \begin{bmatrix} 2\theta \\ \theta^2 \end{bmatrix} (T_{co} - \hat{T}_{co})$$

$$y = \hat{T}_{co} \quad (13)$$

where  $\Phi(\hat{x}) = \Phi(\hat{T}_{co}, \hat{T}_{ho}) \begin{bmatrix} \hat{T}_{co} \\ k_c \Delta \hat{T} \end{bmatrix}$

It should be noted that  $\frac{\partial \Phi(\hat{x})}{\partial \hat{x}}$  must be computed by taking the two possible values of  $\Delta \hat{T}$  into account, as shown in (3). The process model in (12) and the observer in (13) were simulated using the values given in Table 1. The inlet temperatures were considered to be constant  $T_{ci} = 300.5K, T_{hi} = 339.05K$ . The inlet flow rates were  $v_c = 5 \times 10^{-6} m^3 / s, v_h = 1.66 \times 10^{-5} m^3 / s$ .

The simulation was performed using the following initial conditions:  $T_{co}^0 = 315\text{ K}$ ,  $T_{ho}^0 = 333\text{ K}$ ,  $\hat{T}_{co}^0 = 310\text{ K}$  and  $\hat{T}_{ho}^0 = 338\text{ K}$ . The integration step (Euler's numerical method) was  $T_s = 1s$ .

Figures 5 and 6 show the simulation results. It can be seen that both the process output  $T_{co}$  and temperature  $T_{ho}$  are adequately estimated by the observer. The convergence time is about 10 min in both cases. These simulation results were achieved by tuning the observer parameter  $\theta = 0.15$ . Higher values of  $\theta$  guarantee a small convergence time; however, the noise sensitivity of the observer increases. Conversely, smaller values of  $\theta$  reduce the noise sensitivity, but the convergence time increases.

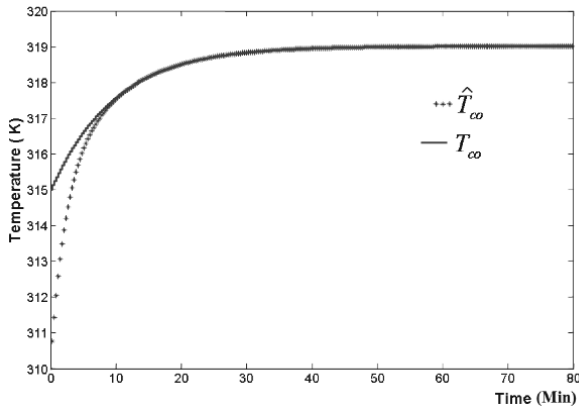


Figure 5. The process output  $T_{co}$  and its estimated value

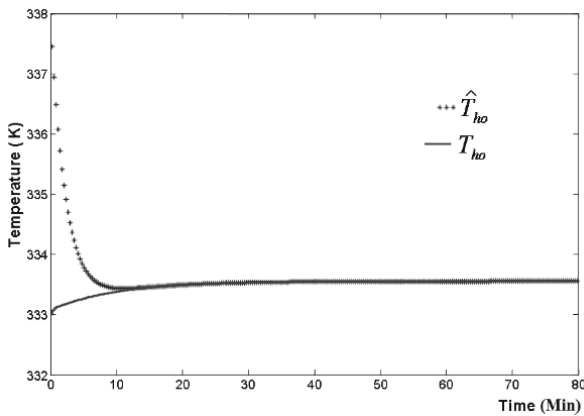


Figure 6.  $T_{ho}$  and its estimated value

Table 1. Physical data used in the simulation and the experiments

Constant	Value	Units
A	$14 \times 10^{-2}$	$m^2$
$\rho_h$	983.3	$kg/m^2$
$\rho_c$	991.8	$kg/m^2$
$V_h$	$15.5 \times 10^{-6}$	$m^3$
$Cp_h$	4179	$J/(K\text{ kg})$
$Cp_c$	4179	$J/(K\text{ kg})$
U	1050	$J/(K\text{ kg})$

### 6. STEP 4: EXPERIMENTAL VALIDATION

The following steps were used to test the observer interacting on-line with the heat exchanger described in Section 2. A monitoring interface was designed in order to read the sensor outputs through a data acquisition board and to display the measured and estimated values of temperatures. The first problem encountered was that the acquisition card Fig.7 (A) acquires voltage sample data whereas the signals provided by the transmitters were electrical currents. This problem was solved by implementing a simple voltage-to-current converter, depicted in Fig.7 (B). It is briefly described in the following section.

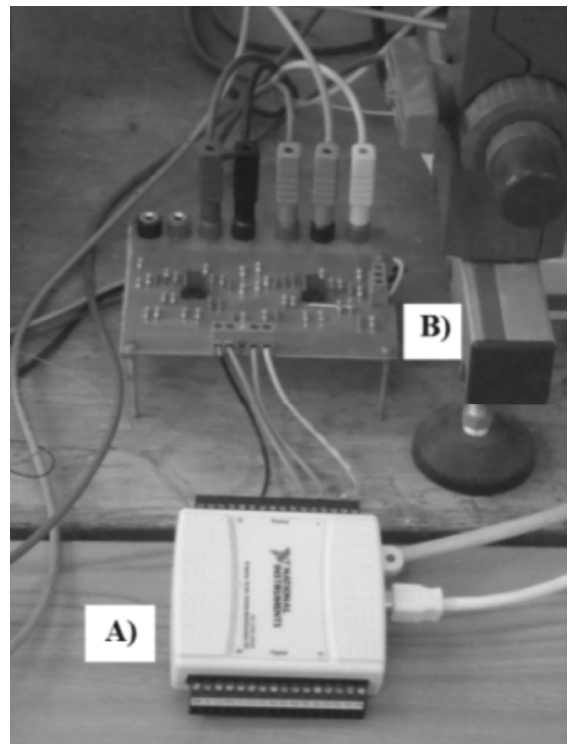


Figure 7. Acquisition card and signal conditioner for the temperature sensor output

## 6.1 The signal conditioner

The temperature sensor has an output current range  $I_s$  of 4-20 mA. The actual current value can be calculated by using the next expression

$$I_s = \frac{1}{1000} \left( 4 + \frac{16}{100} T \right)$$

where  $T$  is the temperature in  $^{\circ}\text{C}$  (the circuit design assumes degrees Celsius and then the conversion to degrees Kelvin is made in the program of the interface). The circuit depicted in Fig. 7 shows four operational amplifiers (OAs). The first one transforms the current  $I_s$  into the voltage  $v_I = -680I_s$ .

The second one is configured like a unitary gain inverting amplifier, consequently, the output voltage of this OA is  $-v_s$ . Finally, the third OA, is a unity-gain inverting amplifier with an adjustable off-set voltage  $v_a$  ( $-12 \leq v_a \leq 0$ ), where  $v_{out} = v_a - v_I$ . As a result, the output voltage of the signal conditioner is

$$v_{out} = \frac{2512}{1000} v_a + \frac{10048}{1000000} T,$$

if  $v_a$  is selected such that it cancels the offset associated to the sensor, i.e  $v_a = 2512 / 1000 = 2.512V$ , then

$$v_{out} = \frac{10048}{1000000} T \cong \frac{1}{10} T.$$

## 6.2 The user interface

The interface was developed using *LabVIEW*<sup>®</sup> in order to provide on-line information display for the supervising engineer in charge of the heat exchanger operation. This monitoring interface performs the following actions (see Fig. 8).

- It displays the process variables graphically (using dotted lines) or numerically. The sensor outputs were sampled through a data acquisition card (AT-MIO-16E-1 from National Instruments).
- It displays graphically (using solid lines) the estimated values computed by the observers.
- It allows changing the tuning parameter of the observer and the sampling time.
- It stores the process variables.

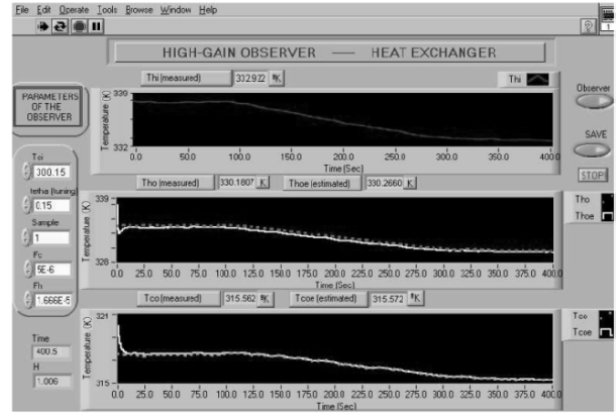


Figure 8. Operator interface

## 6.3 Experimental results

For the experiment described below, the constants and physical data used for the internal model of the observer are given in Table 1. The inlet temperatures were  $T_{ci} = 300.5 \text{ K}$  and  $T_{hi} = 339.15 \text{ K}$ . The inlet flow rates were  $v_c = 5 \times 10^{-6} \text{ m}^3 / \text{s}$  and  $v_h$  had a time-varying profile between  $1.5 \times 10^{-5} \text{ m}^3 / \text{s}$  and  $2 \times 10^{-5} \text{ m}^3 / \text{s}$  as shown in Fig. 9. The initial conditions of the experiment were  $T_{co}^0 = 321 \text{ K}$ ,  $T_{ho}^0 = 335 \text{ K}$ ,  $\hat{T}_{co}^0 = 323 \text{ K}$ , and  $\hat{T}_{ho}^0 = 337 \text{ K}$ .

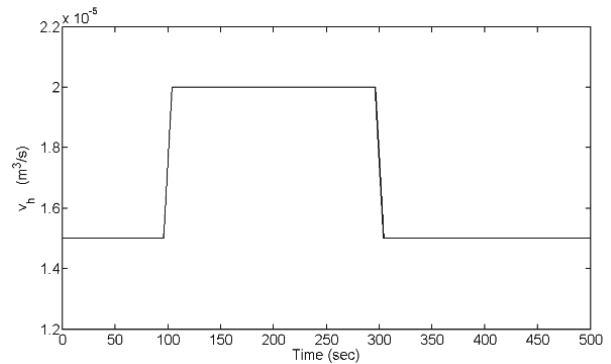
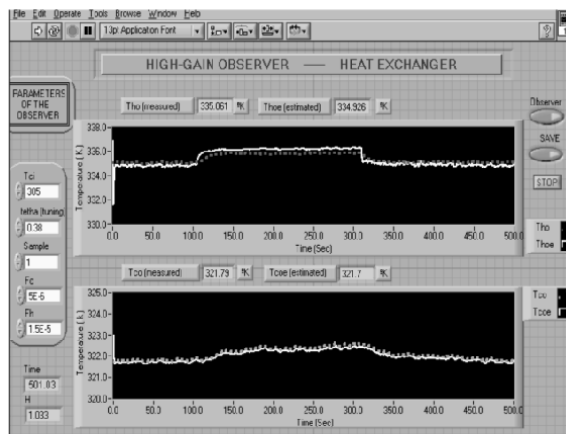


Figure 9. The inlet flow  $v_h$

Temperatures were sampled at  $T_s = 1 \text{ s}$  intervals. The tuning parameter of the observer was  $\theta = 0.38$ . For this experiment, the SIKA<sup>®</sup> glass thermometer TI2 was replaced by a temperature transmitter in order to acquire the corresponding temperature  $T_{ho}$  on-line and to validate the high-gain observer. The results are displayed in Fig. 10. In this figure, the dotted curves correspond to the measured temperatures and the solid curves correspond to the estimates obtained using the high-gain observer. It can be seen that the convergence

time of the estimates  $\hat{T}_{co}, \hat{T}_{ho}$  is very fast (about 5s). Moreover, once the observer converges, it follows the experimental values closely, in spite of the inlet flow rate variations. Although the observer does not yield perfect values of  $T_{ho}$  in the interval of time from  $t = 100$ s to  $t = 300$ s (see the top graph in Fig. 10), these estimated values are acceptable knowing that this variable might depend on the uncertainties of the process related to the lumped parameter  $U$ , which is considered to be constant in this study. An improved version of this observer could be designed if this parameter were considered time-varying.



**Figure 10.** Real  $T_{co}$ ,  $T_{ho}$  (dotted lines) and their estimated values (solid means)

## 7. CONCLUSIONS

The aim of this paper is to contribute to the area of education control, particularly in the field of process monitoring. Initially, this paper briefly explains what a soft sensor is, and then its utility is illustrated via experimental measurements in a heat-exchanger pilot plant. A concentric tube, counter-flow heat exchanger, constructed by Didatec Technologies is used for this purpose. The operation and instrumentation setup of the equipment has been concisely described. A simple model for a counter-flow double-pipe heat exchanger is used to implement the soft sensor [13-15]. This model is sufficient for control objectives, which is the case of observer applications. A high-gain observer type was preferred because of the straightforward calibration and implementation associated procedures (only one parameter) and the satisfactory obtainable robustness (the solution of a Lyapunov function).

The experimental setup includes temperature sensors

to validate the theoretical observer estimations. It has been necessary to adjust the output signals of sensors from current to voltage, before the acquisition with a National Instruments card.

A LabVIEW® user interface was also developed to plot signals and to implement the observer algorithm. Theoretical results were verified by a series of simulations.  $T_{co}$  and  $T_{ho}$  were estimated and confronted with the available experimental values. Comparison results clearly show a suitable performance of the developed soft sensor and offer quite a lot knowledge about the parameter  $\theta$  required for observer estimation: a greater value of  $\theta$  denotes a faster convergence, but a higher noise sensibility as well. Lower  $\theta$  values produce the inverse effect.

The observer estimations evidently agree with simulation calculations and even more so with experimental on-line measurements carried out as a complementary validation test. This assertion can be corroborated analyzing Fig. 10, which is obtained under adverse operation conditions, because inlet flow rate perturbations at the process input were applied to prove the robustness of the estimation.

The results described demonstrate that soft sensors are advantageous when a non-measurable variable is needed to characterize the process performance or when the soft sensors represent an effective substitution for costly sensors. Furthermore, the straightforward recalibration and distance monitoring capabilities make the soft sensors a tool for potentially improving a process.

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